The MCP1726 is a 1A Low Dropout (LDO) linear regulator that provides high current and low output voltages in a very small package. The MCP1726 comes in a fixed (or adjustable) output voltage version, with an output voltage range of 0.8V to 5.0V. The 1A output current capability, combined with the low output voltage capability, make the MCP1726 a good choice for new sub-1.8V output voltage LDO applications that have high current demands.

The MCP1726 is stable using ceramic output capacitors that inherently provide lower output noise and reduce the size and cost of the entire regulator solution. Only 1 µF of output capacitance is needed to stabilize the LDO.

Using CMOS construction, the quiescent current consumed by the MCP1726 is typically less than 140 µA over the entire input voltage range, making it attractive for portable computing applications that demand high output current. When shut down, the quiescent current is reduced to less than 0.1 µA.

The scaled-down output voltage is internally monitored and a power good (PWRGD) output is provided when the output is within 92% of regulation (typical). An external capacitor can be used on the C_DELAY pin to adjust the delay from 1 ms to 300 ms.

The overtemperature and short circuit current-limiting provide additional protection for the LDO during system fault conditions.

### Features
- 1A Output Current Capability
- Input Operating Voltage Range: 2.3V to 6.0V
- Adjustable Output Voltage Range: 0.8V to 5.0V
- Standard Fixed Output Voltages:
  - 0.8V, 1.2V, 1.8V, 2.5V, 3.3V, 5.0V
- Low Dropout Voltage: 220 mV Typical at 1A
- Typical Output Voltage Tolerance: 0.4%
- Stable with 1.0 µF Ceramic Output Capacitor
- Fast response to Load Transients
- Low Supply Current: 140 µA (typ)
- Low Shutdown Supply Current: 0.1 µA (typ)
- Adjustable Delay on Power Good Output
- Short Circuit Current Limiting and Overtemperature Protection
- 3x3 DFN-8 and SOIC-8 Package Options

### Applications
- High-Speed Driver Chipset Power
- Networking Backplane Cards
- Notebook Computers
- Network Interface Cards
- Palmtop Computers
- 2.5V to 1.XV Regulators

### Description

1A, Low Voltage, Low Quiescent Current LDO Regulator

---

### Package Types

<table>
<thead>
<tr>
<th>Adjustable (SOIC-8)</th>
<th>Fixed (SOIC-8)</th>
<th>Adjustable (3x3 DFN)</th>
<th>Fixed (3x3 DFN)</th>
</tr>
</thead>
<tbody>
<tr>
<td>VIN 1&lt;br&gt;VIN 2&lt;br&gt;SHDN 3&lt;br&gt;GND 4</td>
<td>V_IN 1&lt;br&gt;V_OUT 2&lt;br&gt;C_DELAY 3&lt;br&gt;PWRGD 4</td>
<td>V_IN 1&lt;br&gt;V_OUT 2&lt;br&gt;C_DELAY 3&lt;br&gt;PWRGD 4</td>
<td>V_IN 1&lt;br&gt;V_OUT 2&lt;br&gt;C_DELAY 3&lt;br&gt;PWRGD 4</td>
</tr>
<tr>
<td>SHDN 5&lt;br&gt;GND 6</td>
<td>PWRGD 7</td>
<td>PWRGD 7</td>
<td>PWRGD 7</td>
</tr>
<tr>
<td>ADJ 8</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
Typical Application

MCP1726 Fixed Output Voltage

\[ V_{\text{IN}} = 2.3\text{V to 2.8V} \]

- \( C_1 = 4.7\ \mu\text{F} \)
- \( R_1 = 100\ \text{k}\Omega \)
- \( R_2 = 20\ \text{k}\Omega \)
- \( R_3 = 100\ \text{k}\Omega \)

\[ V_{\text{OUT}} = 1.8\text{V @ 1A} \]

MCP1726 Adjustable Output Voltage

\[ V_{\text{IN}} = 2.3\text{V to 2.8V} \]

- \( C_1 = 4.7\ \mu\text{F} \)
- \( C_3 = 1000\ \text{pF} \)

\[ V_{\text{OUT}} = 1.2\text{V @ 1A} \]
1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

VIN..................................................................................................................6.5V
Maximum Voltage on Any Pin .. (GND – 0.3V) to (VDD + 0.3)V
Maximum Junction Temperature, TJ ..................................................+150°C
Maximum Power Dissipation......... Internally-Limited (Note 6)
Storage temperature .....................................-65°C to +150°C

Notice: Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC CHARACTERISTICS

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input Operating Voltage</td>
<td>VIN</td>
<td>2.3</td>
<td>6.0</td>
<td>V</td>
<td>Note 1</td>
<td></td>
</tr>
<tr>
<td>Input Quiescent Current</td>
<td>IQ</td>
<td>—</td>
<td>140</td>
<td>220</td>
<td>µA</td>
<td></td>
</tr>
<tr>
<td>Input Quiescent Current for SHDN Mode</td>
<td>ISHDN</td>
<td>—</td>
<td>0.1</td>
<td>3</td>
<td>µA</td>
<td>SHDN = GND</td>
</tr>
<tr>
<td>Maximum Output Current</td>
<td>IOUT</td>
<td>1</td>
<td>—</td>
<td>—</td>
<td>A</td>
<td>VOUT = 2.3V to 6.0V (Note 1)</td>
</tr>
<tr>
<td>Line Regulation</td>
<td>ΔVOUT/ (VOUT × ΔVIN)</td>
<td>—</td>
<td>0.05</td>
<td>0.3</td>
<td>%/V</td>
<td>(VR + 0.5)V ≤ VIN ≤ 6V</td>
</tr>
<tr>
<td>Load Regulation</td>
<td>ΔVOUT/VOUT</td>
<td>-1.5</td>
<td>±0.5</td>
<td>1.5</td>
<td>%</td>
<td>IOUT = 1 mA to 1A, VIN = (VR + 0.6)V (Note 4)</td>
</tr>
<tr>
<td>Output Short Circuit Current</td>
<td>IOUT_SC</td>
<td>—</td>
<td>1.7</td>
<td>—</td>
<td>A</td>
<td>VIN = (VR + 0.5)V, RLOAD &lt; 0.1Ω, Peak Current</td>
</tr>
</tbody>
</table>

Adjust Pin Characteristics

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Adjust Pin Reference Voltage</td>
<td>VADJ</td>
<td>0.402</td>
<td>0.410</td>
<td>0.418</td>
<td>V</td>
<td>VIN = 2.3V to VIN = 6.0V, IOUT = 1 mA</td>
</tr>
<tr>
<td>Adjust Pin Leakage Current</td>
<td>IADJ</td>
<td>-10</td>
<td>±0.01</td>
<td>+10</td>
<td>nA</td>
<td></td>
</tr>
<tr>
<td>Adjust Temperature Coefficient</td>
<td>TCVOUT</td>
<td>—</td>
<td>40</td>
<td>—</td>
<td>ppm/°C</td>
<td></td>
</tr>
</tbody>
</table>

Fixed-Output Characteristics

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Voltage Regulation</td>
<td>VOUT</td>
<td>VR ≥ 2.5%</td>
<td>VR ±0.5%</td>
<td>VR + 2.5%</td>
<td>V</td>
<td></td>
</tr>
</tbody>
</table>

Dropout Characteristics

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Dropout Voltage</td>
<td>VIN-VOUT</td>
<td>220</td>
<td>—</td>
<td>500</td>
<td>mV</td>
<td>IOUT = 1A, VIN(MIN) = 2.3V (Note 5)</td>
</tr>
</tbody>
</table>

Note 1: The minimum VIN must meet two conditions: VIN ≥ 2.3V and VIN ≥ (VR + 2.5%) + VDROPOUT.

2: VR is the nominal regulator output voltage for the fixed cases. VR = 1.2V, 1.8V, etc. VR is the desired set point output voltage for the adjustable cases. VR = VADJ ± (((VR/VR) + 1) Figure 4-1. Note 5. |

3: TCVOUT = (VOUT-HIGH – VOUT-LOW) × 10^6 / (VR × ΔTemperature). VOUT-HIGH is the highest voltage measured over the temperature range. VOUT-LOW is the lowest voltage measured over the temperature range.

4: Load regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.

5: Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of VIN = VR + 0.5V.

6: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air. (i.e., θJA, θJA). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum 150°C rating. Sustained junction temperatures above 125°C can impact device reliability.

7: The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired junction temperature. The test time is small enough such that the rise in the junction temperature over the ambient temperature is not significant.
## DC CHARACTERISTICS (CONTINUED)

### Electrical Specifications:

- Unless otherwise noted, $V_{IN} = (V_R + 0.5V)$ or 2.3V, whichever is greater, $I_{OUT} = 1 mA$, $C_{IN} = C_{OUT} = 4.7 \mu F$ (X7R Ceramic), $T_A = +25°C$. **Boldface** type applies for junction temperatures, $T_J$ ([Note 7](#)) of -40°C to +125°C.

### Power Good Characteristics

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input Voltage Operating Range for valid PWRGD</td>
<td>$V_{PWRGD_VIN}$</td>
<td>1.0</td>
<td>—</td>
<td>6.0</td>
<td>V</td>
<td>$T_A = +25°C$</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1.2</td>
<td>—</td>
<td>6.0</td>
<td>V</td>
<td>$T_A = -40°C$ to +125°C</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>$I_{SINK} = 100 \mu A$</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PWRGD Threshold Voltage (Referenced to $V_{OUT}$)</td>
<td>PWRGD_THF</td>
<td>88</td>
<td>92</td>
<td>96</td>
<td>% $V_{OUT} &lt; 2.5V$, Falling Edge</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>89</td>
<td>92</td>
<td>95</td>
<td>% $V_{OUT} &gt; 2.5V$, Falling Edge</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>$V_{OUT} &lt; 2.5V$, Rising Edge</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>90</td>
<td>% $V_{OUT} &gt; 2.5V$, Rising Edge</td>
<td></td>
</tr>
<tr>
<td>PWRGD Output Voltage Low</td>
<td>$V_{PWRGD_L}$</td>
<td>—</td>
<td>0.2</td>
<td>0.4</td>
<td>V</td>
<td>$I_{PWRGD_SINK} = 1.2 mA$</td>
</tr>
<tr>
<td>PWRGD Leakage</td>
<td>$P_{PWRGD_L}$</td>
<td>—</td>
<td>0.1</td>
<td>—</td>
<td>µA</td>
<td>$V_{PWRGD} = V_{IN} = 6.0V$</td>
</tr>
<tr>
<td>PWRGD Time Delay</td>
<td>$T_{PG}$</td>
<td></td>
<td>10</td>
<td>55</td>
<td>ms</td>
<td>$C_{DELAY} = OPEN$</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>30</td>
<td></td>
<td>ms</td>
<td>$C_{DELAY} = 0.01 \mu F$</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>$C_{DELAY} = 0.1 \mu F$</td>
</tr>
<tr>
<td>Detect Threshold to PWRGD Active Time Delay</td>
<td>$T_{VDET_PWRGD}$</td>
<td>—</td>
<td>170</td>
<td>—</td>
<td>µs</td>
<td></td>
</tr>
</tbody>
</table>

### Shutdown Input

- Logic-High Input $V_{SHDN\_HIGH}$ | 45 | %$V_{IN}$ | $V_{IN} = 2.3V$ to 6.0V |
- Logic-Low Input $V_{SHDN\_LOW}$ | 15 | %$V_{IN}$ | $V_{IN} = 2.3V$ to 6.0V |
- SHDN Input Leakage Current $SHDN_{ILK}$ | -0.1 | ±0.001 | +0.1 | µA | $V_{IN} = 6V$, SHDN = GND, SHDN = GND |

### AC Performance

- Output Delay From SHDN $T_{OR}$ | 100 | µs | SHDN = GND to $V_{IN}$ |
- Output Noise $e_{N}$ | 2.0 | — | $I_{OUT} = 200 mA$, $f = 1 kHz$, $C_{OUT} = 1 \mu F$ (X7R Ceramic), $V_{OUT} = 2.5V$ |
- Power Supply Ripple Rejection Ratio $PSRR$ | 54 | — | dB | $f = 100 Hz$, $C_{OUT} = 10 \mu F$, $I_{OUT} = 100 mA$, $V_{INAC} = 30 mV$ pk-pk, $C_{IN} = 0 \mu F$ |
- Thermal Shutdown Temperature $T_{SD}$ | 150 | — | °C | $I_{OUT} = 100 \mu A$, $V_{OUT} = 1.8V$, $V_{IN} = 2.8V$ |
- Thermal Shutdown Hysteresis $\Delta T_{SD}$ | 10 | — | °C | $I_{OUT} = 100 \mu A$, $V_{OUT} = 1.8V$, $V_{IN} = 2.8V$ |

### Notes

1. The minimum $V_{IN}$ must meet two conditions: $V_{IN} \geq 2.3V$ and $V_{IN} \geq (VR + 2.5\%) + V_{DROPOUT}$.
2. $V_R$ is the nominal regulator output voltage for the fixed cases. $V_R = 1.2V$, 1.8V, etc. $V_R$ is the desired set point output voltage for the adjustable cases. $V_R = V_{ADJ} * ((R_1/R_2)+1)$. **Figure 4-1**.
3. $TC_{OUT} = (V_{OUT\_HIGH} - V_{OUT\_LOW}) * 10^6 / (V_R * \Delta Temperature)$, $V_{OUT\_HIGH}$ is the highest voltage measured over the temperature range, $V_{OUT\_LOW}$ is the lowest voltage measured over the temperature range.
4. Load regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.
5. Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of $V_{IN} = V_R + 0.5V$.
6. The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., $T_A$, $T_J$, $R_{JA}$). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum 150°C rating. Sustained junction temperatures above 125°C can impact device reliability.
7. The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired junction temperature. The test time is small enough such that the rise in the junction temperature over the ambient temperature is not significant.
### TEMPERATURE SPECIFICATIONS

**Electrical Specifications:** Unless otherwise indicated, all limits apply for $V_{\text{IN}} = 2.3\,\text{V to 6.0V}$. 

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Temperature Ranges</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Operating Junction Temperature Range</td>
<td>$T_J$</td>
<td>-40</td>
<td></td>
<td>+125</td>
<td>°C</td>
<td>Steady State</td>
</tr>
<tr>
<td>Maximum Junction Temperature</td>
<td>$T_J$</td>
<td>—</td>
<td>—</td>
<td>+150</td>
<td>°C</td>
<td>Transient</td>
</tr>
<tr>
<td>Storage Temperature Range</td>
<td>$T_A$</td>
<td>-65</td>
<td>—</td>
<td>+150</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td><strong>Thermal Package Resistances</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Thermal Resistance, 8LD 3x3 DFN</td>
<td>$\theta_{JA}$</td>
<td>—</td>
<td>41</td>
<td>—</td>
<td>°C/W</td>
<td>4-Layer JC51-7 Standard Board with vias</td>
</tr>
<tr>
<td>Thermal Resistance, 8LD SOIC</td>
<td>$\theta_{JA}$</td>
<td>—</td>
<td>150</td>
<td>—</td>
<td>°C/W</td>
<td>4-Layer JC51-7 Standard Board</td>
</tr>
</tbody>
</table>
2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

NOTE: Unless otherwise indicated, \( V_{IN} = V_{OUT} + 0.5V \), \( I_{OUT} = 1\ mA \) and \( T_A = +25°C \).

**FIGURE 2-1:** Quiescent Current vs. Input Voltage (1.2V Adjustable).

**FIGURE 2-2:** Ground Current vs. Load Current (1.2V Adjustable).

**FIGURE 2-3:** Quiescent Current vs. Junction Temperature (1.2V Adjustable).

**FIGURE 2-4:** Line Regulation vs. Temperature (1.2V Adjustable).

**FIGURE 2-5:** Load Regulation vs. Temperature.

**FIGURE 2-6:** Adjust Pin Voltage vs. Temperature.
NOTE: Unless otherwise indicated, \( V_{\text{IN}} = V_{\text{OUT}} + 0.5V \), \( I_{\text{OUT}} = 1\) mA and \( T_A = +25^\circ\text{C} \).

**FIGURE 2-7:** Dropout Voltage vs. Output Current (Adjustable Version).

**FIGURE 2-8:** Dropout Voltage vs. Temperature (Adjustable Version).

**FIGURE 2-9:** Power Good (PWRGD) Time Delay vs. Temperature.

**FIGURE 2-10:** Quiescent Current vs. Input Voltage (0.8V Fixed).

**FIGURE 2-11:** Quiescent Current vs. Input Voltage (3.3V Fixed).

**FIGURE 2-12:** Ground Current vs. Load Current.
NOTE: Unless otherwise indicated, \( V_{IN} = V_{OUT} + 0.5V \), \( I_{OUT} = 1 \) mA and \( T_A = +25°C \).

**FIGURE 2-13:** Quiescent Current vs. Temperature.

**FIGURE 2-14:** \( I_{SHDN} \) vs. Temperature.

**FIGURE 2-15:** Line Regulation vs. Temperature (0.8V Fixed)

**FIGURE 2-16:** Line Regulation vs. Temperature (3.3V Fixed).

**FIGURE 2-17:** Load Regulation vs. Temperature (\( V_{OUT} < 2.5V \) Fixed).

**FIGURE 2-18:** Load Regulation vs. Temperature (\( V_{OUT} \geq 2.5V \) Fixed).
NOTE: Unless otherwise indicated, $V_{\text{IN}} = V_{\text{OUT}} + 0.5V$, $I_{\text{OUT}} = 1\ mA$ and $T_A = +25^\circ C$.

**FIGURE 2-19:** Dropout Voltage vs. Load Current.

**FIGURE 2-20:** Dropout Voltage vs. Temperature.

**FIGURE 2-21:** Short Circuit Current vs. Input Voltage.

**FIGURE 2-22:** Output Noise Voltage Density vs. Frequency.

**FIGURE 2-23:** Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{\text{OUT}} = 1.2\ V$ Adj.).

**FIGURE 2-24:** Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{\text{OUT}} = 1.2\ V$ Adj.).
NOTE: Unless otherwise indicated, $V_{IN} = V_{OUT} + 0.5\text{V}$, $I_{OUT} = 1\text{ mA}$ and $T_A = +25^\circ\text{C}$.

**FIGURE 2-25:** Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{OUT} = 2.5\text{V}$ Fixed).

**FIGURE 2-26:** Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{OUT} = 2.5\text{V}$ Fixed).

**FIGURE 2-27:** 2.5V (Adj.) Startup from $V_{IN}$.

**FIGURE 2-28:** 2.5V (Adj.) Startup from Shutdown.

**FIGURE 2-29:** Power Good (PWRGD) Timing with $C_{BYPASS}$ of 1000 pF.

**FIGURE 2-30:** Power Good (PWRGD) Timing with $C_{BYPASS}$ of 0.01 µF.
NOTE: Unless otherwise indicated, $V_{\text{IN}} = V_{\text{OUT}} + 0.5V$, $I_{\text{OUT}} = 1 \text{ mA}$ and $T_A = +25^\circ \text{C}$.

**FIGURE 2-31:** Dynamic Line Response
(1.2V Fixed).

**FIGURE 2-32:** Dynamic Line Response
(2.5V Fixed).

**FIGURE 2-33:** Dynamic Load Response
(2.5V Fixed, 10 mA to 1000 mA).

**FIGURE 2-34:** Dynamic Load Response
(2.5V Fixed, 100 mA to 1000 mA).
3.0 PIN DESCRIPTION

The descriptions of the pins are listed in Table 3-1.

### Table 3-1: PIN FUNCTION TABLE

<table>
<thead>
<tr>
<th>Pin No. Fixed Output</th>
<th>Pin No. Adjustable Output</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1</td>
<td>V\text{IN}</td>
<td>Input Voltage Supply</td>
</tr>
<tr>
<td>2</td>
<td>2</td>
<td>V\text{IN}</td>
<td>Input Voltage Supply</td>
</tr>
<tr>
<td>3</td>
<td>3</td>
<td>SHDN</td>
<td>Shutdown Control Input (active-low)</td>
</tr>
<tr>
<td>4</td>
<td>4</td>
<td>GND</td>
<td>Ground</td>
</tr>
<tr>
<td>5</td>
<td>5</td>
<td>PWRGD</td>
<td>Power Good Output</td>
</tr>
<tr>
<td>6</td>
<td>6</td>
<td>C\text{DELAY}</td>
<td>Power Good Delay Set-Point Input</td>
</tr>
<tr>
<td>7</td>
<td>7</td>
<td>ADJ</td>
<td>Output Voltage Sense Input (adjustable version)</td>
</tr>
<tr>
<td>8</td>
<td>8</td>
<td>V\text{OUT}</td>
<td>Regulated Output Voltage</td>
</tr>
<tr>
<td>Exposed Pad</td>
<td>Exposed Pad</td>
<td>EP</td>
<td>Exposed Pad of the DFN Package</td>
</tr>
</tbody>
</table>

3.1 Input Voltage Supply (V\text{IN})

Connect the unregulated or regulated input voltage source to V\text{IN}. If the input voltage source is located several inches away from the LDO, or the input source is a battery, it is recommended that an input capacitor be used. A typical input capacitance value of 1 µF to 10 µF should be sufficient for most applications.

3.2 Shutdown Control Input (SHDN)

The SHDN input is used to turn the LDO output voltage on and off. When the SHDN input is at a logic-high level, the LDO output voltage is enabled. When the SHDN input is pulled to a logic-low level, the LDO output voltage is disabled. When the SHDN input is pulled low, the PWRGD output also goes low and the LDO enters a low quiescent current shutdown state where the typical quiescent current is 0.1 µA.

3.3 Ground (GND)

Connect the GND pin of the LDO to a quiet circuit ground. This will help the LDO power supply rejection ratio and noise performance. The ground pin of the LDO only conducts the quiescent current of the LDO (typically 140 µA), so a heavy trace is not required.

3.4 Power Good Output (PWRGD)

The PWRGD output is an open-drain output used to indicate when the LDO output voltage is within 92% (typically) of its nominal regulation value. The PWRGD output has a typical hysteresis value of 2% for the adjustable voltage version and for voltage outputs less than 2.5V. For fixed output voltage versions greater than 2.5V, the hysteresis is 0.7%. The PWRGD output is delayed on power-up by 200 µs (typical, no capacitance on C\text{DELAY} pin). This delay time is controlled by the C\text{DELAY} pin.

3.5 Power Good Delay Set-Point Input (C\text{DELAY})

The C\text{DELAY} input sets the power-up delay time for the PWRGD output. By connecting an external capacitor from the C\text{DELAY} pin to ground, the delay times for the PWRGD output can be adjusted from 200 µs (no capacitance) to 300 ms (0.1 µF capacitor). This allows for the optimal setting of the system reset time.

3.6 Output Voltage Sense Input (ADJ)

The output voltage adjust pin (ADJ) for the adjustable output voltage version of the MCP1726 allows the user to set the output voltage of the LDO by using two external resistors. The adjust pin voltage is 0.41V (typical).

3.7 Regulated Output Voltage (V\text{OUT})

The V\text{OUT} pin(s) is the regulated output voltage of the LDO. A minimum output capacitance of 1.0 µF is required for LDO stability. The MCP1726 is stable with ceramic, tantalum and aluminum-electrolytic capacitors. See Section 4.3 “Output Capacitor” for output capacitor selection guidance.

3.8 Exposed Pad (EP)

The 3x3 DFN package has an exposed pad on the bottom of the package. This pad should be soldered to the Printed Circuit Board (PCB) to aid in the removal of heat from the package during operation. The exposed pad is at the ground potential of the LDO.
4.0 DEVICE OVERVIEW

The MCP1726 is a high output current, Low Dropout (LDO) voltage regulator with an adjustable delay power-good output and shutdown control input. The low dropout voltage of 220 mV at 1A of current makes it ideal for battery-powered applications. Unlike other high output current LDOs, the MCP1726 only draws 220 µA of quiescent current at full load.

4.1 LDO Output Voltage

The MCP1726 LDO is available with either a fixed output voltage or an adjustable output voltage. The output voltage range is 0.8V to 5.5V for both versions.

4.1.1 ADJUST INPUT

The adjustable version of the MCP1726 uses the ADJ pin (pin 7) to get the output voltage feedback for output voltage regulation. This allows the user to set the output voltage of the device with two external resistors. The nominal voltage for ADJ is 0.41V.

Figure 4-1 shows the adjustable version of the MCP1726. Resistors R1 and R2 form the resistor divider network necessary to set the output voltage. With this configuration, the equation for setting VOUT is:

\[
V_{\text{OUT}} = V_{\text{ADJ}} \left(\frac{R_1 + R_2}{R_2}\right)
\]

**EQUATION 4-1:**

\[
V_{\text{OUT}} = \text{LDO Output Voltage}
\]

\[
V_{\text{ADJ}} = \text{ADJ Pin Voltage (typically 0.41V)}
\]

**FIGURE 4-1:** Typical adjustable output voltage application circuit.

The allowable resistance value range for resistor R2 is from 10 kΩ to 200 kΩ. Solving the equation for R1 yields the following equation:

\[
R_1 = R_2 \left(\frac{V_{\text{OUT}} - V_{\text{ADJ}}}{V_{\text{ADJ}}}\right)
\]

**EQUATION 4-2:**

\[
V_{\text{OUT}} = \text{LDO Output Voltage}
\]

\[
V_{\text{ADJ}} = \text{ADJ Pin Voltage (typically 0.41V)}
\]

4.2 Output Current and Current Limiting

The MCP1726 LDO is tested and ensured to supply a minimum of 1A of output current. The MCP1726 has no minimum output load, so the output load current can go to 0 mA and the LDO will continue to regulate the output voltage to within tolerance.

The MCP1726 also incorporates an output current limit. If the output voltage falls below 0.7V due to an overload condition (usually represents a shorted load condition), the output current is limited to 1.7A (typical). If the overload condition is a soft overload, the MCP1726 will supply higher load currents of up to 3A. The MCP1726 should not be operated in this condition continuously as it may result in failure of the device. However, this does allow for device usage in applications that have higher pulsed load currents having an average output current value of 1A or less.

Output overload conditions may also result in an overtemperature shutdown of the device. If the junction temperature rises above 150°C, the LDO will shut down the output voltage. See Section 4.9 “Overtemperature Protection” for more information on overtemperature shutdown.

4.3 Output Capacitor

The MCP1726 requires a minimum output capacitance of 1 µF for output voltage stability. Ceramic capacitors are recommended because of their size, cost and environmental robustness qualities.

Aluminum-electrolytic and tantalum capacitors can be used on the LDO output as well. The Equivalent Series Resistance (ESR) of the electrolytic output capacitor must be no greater than 2 ohms. The output capacitor should be located as close to the LDO output as is practical. Ceramic materials X7R and X5R have low temperature coefficients and are well within the acceptable ESR range required. A typical 1 µF X7R 0805 capacitor has an ESR of 50 milli-ohms.

Larger LDO output capacitors can be used with the MCP1726 to improve dynamic performance and power supply ripple rejection performance. A maximum of 22 µF is recommended. Aluminum-electrolytic capacitors are not recommended for low-temperature applications of < -25°C.
4.4 Input Capacitor

Low input source impedance is necessary for the LDO output to operate properly. When operating from batteries, or in applications with long lead length (> 10 inches) between the input source and the LDO, some input capacitance is recommended. A minimum of 1.0 µF to 4.7 µF is recommended for most applications.

For applications that have output step load requirements, the input capacitance of the LDO is very important. The input capacitance provides the LDO with a good local low-impedance source to pull the transient currents from in order to respond quickly to the output load step. For good step response performance, the input capacitor should be of equivalent (or higher) value than the output capacitor. The capacitor should be placed as close to the input of the LDO as is practical. Larger input capacitors will also help reduce any high-frequency noise on the input and output of the LDO and reduce the effects of any inductance that exists between the input source voltage and the input capacitance of the LDO.

4.5 Power Good Output (PWRGD)

The PWRGD output is used to indicate when the output voltage of the LDO is within 92% (typical value, see the Electrical Characteristics table for Min/Max specs) of its nominal regulation value.

As the output voltage of the LDO rises, the PWRGD output will be held low until the output voltage has exceeded the power good threshold plus the hysteresis value. Once this threshold has been exceeded, the power good time delay is started (shown as TPG in the Electrical Characteristics table). The power good time delay is adjustable via the CDELAY pin of the LDO (see Section 4.6 “CDELAY Input”). By placing a capacitor from the CDELAY pin to ground, the PWRGD power-up time delay can be adjusted from 200 µs (no capacitance) to 300 ms (0.1 µF capacitor). After the time delay period, the PWRGD output will go high, indicating that the output voltage is stable and within regulation limits.

If the output voltage of the LDO falls below the power good threshold, the power good output will transition low. The power good circuitry has a 170 µs delay when detecting a falling output voltage, which helps to increase noise immunity of the power good output and avoid false triggering of the power good output during fast output transients. See Figure 4-2 for power good timing characteristics.

When the LDO is put into Shutdown mode using the SHDN input, the power good output is pulled low immediately, indicating that the output voltage will be out of regulation. The timing diagram for the power good output when using the shutdown input is shown in Figure 4-3.

The power good output is an open-drain output that can be pulled up to any voltage that is equal to or less than the LDO input voltage. This output is capable of sinking 1.2 mA (VPWRGD < 0.4V maximum).
Once the power good threshold (rising) has been reached, the C_DELAY pin charges the external capacitor to 1.5V (typical, this level can vary between 1.4V and 1.75V across the input voltage range of the part). The PWRGD output will transition high when the C_DELAY pin voltage has charged to 0.42V. If the output falls below the power good threshold limit during the charging time between 0.0V and 0.42V on the C_DELAY pin, the C_DELAY pin voltage will be pulled to ground, thus resetting the timer. The C_DELAY pin will be held low until the output voltage of the LDO has once again risen above the power good rising threshold. A timing diagram showing C_DELAY, PWRGD and VOUT is shown in Figure 4-4.

4.7 Shutdown Input (SHDN)

The SHDN input is an active-low input signal that turns the LDO on and off. The SHDN threshold is a percentage of the input voltage. The typical value of this shutdown threshold is 30% of V_IN, with minimum and maximum limits over the entire operating temperature range of 45% and 15%, respectively.

The SHDN input will ignore low-going pulses (pulses meant to shut down the LDO) that are up to 400 ns in pulse width. If the shutdown input is pulled low for more than 400 ns, the LDO will enter Shutdown mode. This small bit of filtering helps to reject any system noise spikes on the shutdown input signal.

On the rising edge of the SHDN input, the shutdown circuitry has a 30 µs delay before allowing the LDO output to turn on. This delay helps to reject any false turn-on signals or noise on the SHDN input signal. After the 30 µs delay, the LDO output enters its soft-start period as it rises from 0V to its final regulation value. If the SHDN input signal is pulled low during the 30 µs delay period, the timer will be reset and the delay time will start over again on the next rising edge of the SHDN input. The total time from the SHDN input going high (turn-on) to the LDO output being in regulation is typically 100 µs. See Figure 4-5 for a timing diagram of the SHDN input.

Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below the nominal value that was measured with a VR + 0.5V differential applied. The MCP1726 LDO has a very low dropout voltage specification of 220 mV (typical) at 1A of output current. See the Electrical Characteristics table for maximum dropout voltage specifications.

The MCP1726 LDO operates across an input voltage range of 2.3V to 6.0V and incorporates input Undervoltage Lockout (UVLO) circuitry that keeps the LDO output voltage off until the input voltage reaches a minimum of 2.18V (typical) on the rising edge of the input voltage. As the input voltage falls, the LDO output will remain on until the input voltage level reaches 2.04V (typical).

Since the MCP1726 LDO undervoltage lockout activates at 2.04V as the input voltage is falling, the dropout voltage specification does not apply for output voltages that are less than 1.9V.

For high-current applications, voltage drops across the PCB traces must be taken into account. The trace resistances can cause significant voltage drops between the input voltage source and the LDO. For applications with input voltages near 2.3V, these PCB trace voltage drops can sometimes lower the input voltage enough to trigger a shutdown due to undervoltage lockout.
4.9  Overtemperature Protection

The MCP1726 LDO has temperature-sensing circuitry to prevent the junction temperature from exceeding approximately 150°C. If the LDO junction temperature does reach 150°C, the LDO output will be turned off until the junction temperature cools to approximately 140°C, at which point the LDO output will automatically resume normal operation. If the internal power dissipation continues to be excessive, the device will again shut off. The junction temperature of the die is a function of power dissipation, ambient temperature and package thermal resistance. See Section 5.0 “Application Circuits/Issues” for more information on LDO power dissipation and junction temperature.
5.0 APPLICATION CIRCUITS/ ISSUES

5.1 Typical Application
The MCP1726 is used for applications that require high LDO output current and a power good output.

5.1.1 APPLICATION CONDITIONS
- Package Type = 3x3DFN8
- Input Voltage Range = 3.3V ± 10%
- \( V_{IN} \) maximum = 3.63V
- \( V_{IN} \) minimum = 2.97V
- \( V_{OUT} \) typical = 2.5V
- \( I_{OUT} \) = 1.0A maximum

5.2 Power Calculations

5.2.1 POWER DISSIPATION
The internal power dissipation within the MCP1726 is a function of input voltage, output voltage, output current and quiescent current. The following equation can be used to calculate the internal power dissipation for the LDO.

**EQUATION 5-1:**

\[
P_{LDO} = (V_{IN(MAX)} - V_{OUT(MIN)}) \times I_{OUT(MAX)}
\]

- \( P_{LDO} \) = LDO Pass device internal power dissipation
- \( V_{IN(MAX)} \) = Maximum input voltage
- \( V_{OUT(MIN)} \) = LDO minimum output voltage

In addition to the LDO pass element power dissipation, there is power dissipation within the MCP1726 as a result of quiescent or ground current. The power dissipation as a result of the ground current can be calculated using the following equation:

**EQUATION 5-2:**

\[
P_{I(GND)} = V_{IN(MAX)} \times I_{VIN}
\]

- \( P_{I(GND)} \) = Power dissipation due to the quiescent current of the LDO
- \( V_{IN(MAX)} \) = Maximum input voltage
- \( I_{VIN} \) = Current flowing in the \( V_{IN} \) pin with no LDO output current (LDO quiescent current)

The total power dissipated within the MCP1726 is the sum of the power dissipated in the LDO pass device and the \( P_{I(GND)} \) term. Because of the CMOS construction, the typical \( I_{GND} \) for the MCP1726 is 140 µA. Operating at a maximum of 3.63V results in a power dissipation of 0.51 milli-Watts. For most applications, this is small compared to the LDO pass device power dissipation and can be neglected.

The maximum continuous operating junction temperature specified for the MCP1726 is +125°C. To estimate the internal junction temperature of the MCP1726, the total internal power dissipation is multiplied by the thermal resistance from junction to ambient (\( R_{\theta JA} \)) of the device. The thermal resistance from junction to ambient for the 3x3DFN package is estimated at 41°C/W.

**EQUATION 5-3:**

\[
T_{J(MAX)} = P_{TOTAL} \times R_{\theta JA} + T_{AMAX}
\]

- \( T_{J(MAX)} \) = Maximum continuous junction temperature
- \( P_{TOTAL} \) = Total device power dissipation
- \( R_{\theta JA} \) = Thermal resistance from junction-to-ambient
- \( T_{AMAX} \) = Maximum ambient temperature
The maximum power dissipation capability for a package can be calculated given the junction-to-ambient thermal resistance and the maximum ambient temperature for the application. The following equation can be used to determine the package maximum internal power dissipation.

**EQUATION 5-4:**

\[
P_{D(MAX)} = \frac{(T_{J(MAX)} - T_A(MAX))}{R_{θ_{JA}}}
\]

- \(P_{D(MAX)}\) = Maximum device power dissipation
- \(T_{J(MAX)}\) = maximum continuous junction temperature
- \(T_A(MAX)\) = maximum ambient temperature
- \(R_{θ_{JA}}\) = Thermal resistance from junction-to-ambient

**EQUATION 5-5:**

\[
T_{J(RISE)} = P_{D(MAX)} \times R_{θ_{JA}}
\]

- \(T_{J(RISE)}\) = Rise in device junction temperature over the ambient temperature
- \(P_{D(MAX)}\) = Maximum device power dissipation
- \(R_{θ_{JA}}\) = Thermal resistance from junction-to-ambient

**EQUATION 5-6:**

\[
T_{J} = T_{J(RISE)} + T_A
\]

- \(T_J\) = Junction temperature
- \(T_{J(RISE)}\) = Rise in device junction temperature over the ambient temperature
- \(T_A\) = Ambient temperature

### 5.3 Typical Application

Internal power dissipation, junction temperature rise, junction temperature and maximum power dissipation is calculated in the following example. The power dissipation as a result of ground current is small enough to be neglected.

#### 5.3.1 POWER DISSIPATION EXAMPLE

**Package**

Package Type = 3x3DFN

**Input Voltage**

\(V_{IN} = 3.3V \pm 10\%\)

**LDO Output Voltage and Current**

\(V_{OUT} = 2.5V\)

\(I_{OUT} = 1.0A\)

**Maximum Ambient Temperature**

\(T_A(MAX) = 70°C\)

**Internal Power Dissipation**

\(P_{LDO(MAX)} = (V_{IN(MAX)} - V_{OUT(MIN)}) \times I_{OUT(MAX)}\)

\(P_{LDO} = (3.3V \times 1.1) - (0.975 \times 2.5V) \times 1.0A\)

\(P_{LDO} = 1.192\) Watts

**Device Junction Temperature Rise**

The internal junction temperature rise is a function of internal power dissipation and the thermal resistance from junction to ambient for the application. The thermal resistance from junction to ambient (\(R_{θ_{JA}}\)) is derived from an EIA/JEDEC standard for measuring thermal resistance for small surface-mount packages. The EIA/JEDEC specification is JESD51-7 “High Effective Thermal Conductivity Test Board for Leaded Surface-Mount Packages”. The standard describes the test method and board specifications for measuring the thermal resistance from junction to ambient. The actual thermal resistance for a particular application can vary depending on many factors such as copper area and thickness. Refer to AN792, “A Method to Determine How Much Power a SOT23 Can Dissipate in an Application” (DS00792), for more information regarding this subject.

\[
T_{J(RISE)} = P_{TOTAL} \times R_{θ_{JA}}
\]

\(T_{J(RISE)} = 1.192\) W \(\times 41.0°C/W\)

\(T_{J(RISE)} = 48.8°C\)
Junction Temperature Estimate

To estimate the internal junction temperature, the calculated temperature rise is added to the ambient or offset temperature. For this example, the worst-case junction temperature is estimated below:

\[ T_J = T_{\text{RISE}} + T_{A(\text{MAX})} \]
\[ T_J = 48.8°C + 70.0°C \]
\[ T_J = 118.8°C \]

As you can see from the result, this application will be operating very near the maximum operating junction temperature of 125°C. The PCB layout for this application is very important as it has a significant impact on the junction-to-ambient thermal resistance (\( R_{\text{JA}} \)) of the 3x3 DFN package, which is very important in this application.

Maximum Package Power Dissipation at 70°C Ambient Temperature

3x3 DFN (41°C/W \( R_{\text{JA}} \))

\[ P_{D(\text{MAX})} = \frac{(125°C - 70°C)}{41°C/W} \]
\[ P_{D(\text{MAX})} = 1.34W \]

8LD SOIC (150°C/Watt \( R_{\text{JA}} \))

\[ P_{D(\text{MAX})} = \frac{(125°C - 70°C)}{150°C/W} \]
\[ P_{D(\text{MAX})} = 0.366W \]

From this table you can see the difference in maximum allowable power dissipation between the 3x3 DFN package and the 8-pin SOIC package. This difference is due to the exposed metal tab on the bottom of the DFN package. The exposed tab of the DFN package provides a very good thermal path from the die of the LDO to the PCB. The PCB then acts like a heatsink, providing more area to distribute the heat generated by the LDO.
## 6.0 PACKAGING INFORMATION

### 6.1 Package Marking Information

#### 8-Lead SOIC (150 mil) Example:

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<th>XXXY</th>
<th>WW</th>
<th>NNN</th>
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</thead>
<tbody>
<tr>
<td>CAAA</td>
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<td>256</td>
<td></td>
</tr>
</tbody>
</table>

#### 8-Lead DFN (3x3) Example:

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<th>YYYY</th>
<th>WW</th>
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</thead>
<tbody>
<tr>
<td>CAAA</td>
<td>E543</td>
<td>256</td>
<td></td>
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</tbody>
</table>

#### Voltage Option Code

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<td>1.2V</td>
<td>CAAB</td>
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<tr>
<td>1.8V</td>
<td>CAAC</td>
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<tr>
<td>2.5V</td>
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<td>3.0V</td>
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<tr>
<td>3.3V</td>
<td>CAAF</td>
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<td>5.0V</td>
<td>CAAG</td>
</tr>
<tr>
<td>Adj</td>
<td>AADJ</td>
</tr>
</tbody>
</table>

### Legend:

- XX...X  Customer-specific information
- Y       Year code (last digit of calendar year)
- YY      Year code (last 2 digits of calendar year)
- WW      Week code (week of January 1 is week ‘01’)
- NNN     Alphanumeric traceability code
- e3      Pb-free JEDEC designator for Matte Tin (Sn)
- *       This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

### Note:

In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.
8-Lead Plastic Dual Flat, No Lead Package (MF) – 3x3x0.9 mm Body [DFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

<table>
<thead>
<tr>
<th>Units</th>
<th>MILLIMETERS</th>
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<td>Pitch</td>
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<td>Contact Thickness</td>
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<td>Exposed Pad Width</td>
<td>E2</td>
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<td>Overall Width</td>
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<tr>
<td>Exposed Pad Length</td>
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<tr>
<td>Contact Width</td>
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<tr>
<td>Contact Length</td>
<td>L</td>
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<tr>
<td>Contact-to-Exposed Pad</td>
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</tr>
</tbody>
</table>

Notes:
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package may have one or more exposed tie bars at ends.
3. Package is saw singulated.
4. Dimensioning and tolerancing per ASME Y14.5M.
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
   REF: Reference Dimension, usually without tolerance, for information purposes only.
8-Lead Plastic Small Outline (SN) – Narrow, 3.90 mm Body [SOIC]

**Notes:**
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15 mm per side.
4. Dimensioning and tolerancing per ASME Y14.5M.
   
   **BSC:** Basic Dimension. Theoretically exact value shown without tolerances.
   **REF:** Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-057B
APPENDIX A: REVISION HISTORY

Revision C (August 2007)
The following is the list of modifications:
1. Added 3.0V option to Packing Marking section.
2. Updated package outline drawings.
3. Added 3.0V option to Product Identification System (PIS) section.

Revision B (March 2005)
The following is the list of modifications:
1. Replaced 3x3 DFN package diagram.
2. Emphasized (bolded) a few specifications of Section 1.0 “Electrical Characteristics” in the DC Characteristics table.

Revision A (February 2005)
• Original Release of this Document.
# PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

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<th>PART NO.</th>
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<th>XXX</th>
<th>X</th>
<th>X</th>
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<tbody>
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<td>Voltage Output</td>
<td>Tolerance</td>
<td>Temp. Range</td>
<td>Package</td>
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<td>MCP1726:1A</td>
<td>T = Tape and Reel</td>
<td>080 = 0.80V</td>
<td>2 = 2.0%</td>
<td>E = -40°C to +125°C</td>
<td>SN = Plastic SOIC, (150 mil Body) 8-Lead</td>
</tr>
<tr>
<td>Standard Output Voltage *</td>
<td>Blank = Tube</td>
<td>120 = 1.20V</td>
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<td>MF = Plastic Dual Flat No Lead, 3x3 mm Body (DFN), 8-Lead</td>
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<tr>
<td></td>
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<td>180 = 1.80V</td>
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<td>250 = 2.50V</td>
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<td>300 = 3.00V</td>
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<td>500 = 5.00V</td>
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<tr>
<td>ADJ = Adjustable Voltage Version</td>
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</tr>
<tr>
<td>* Custom output voltages available upon request. Contact your local Microchip sales office for more information.</td>
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</tr>
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</table>

## Examples:

- a) MCP1726-0802E/MF: 0.8V, 1A LDO, 8LD DFN Pkg.
- b) MCP1726-1202E/SN: 1.20V, 1A LDO, 8LD SOIC Pkg.
- c) MCP1726T-1802E/MF: Tape and Reel, 1.80V, 1A LDO, 8LD DFN Pkg.
- d) MCP1726-2502E/SN: 2.50V, 1A LDO, 8LD SOIC Pkg.
- e) MCP1726-3302E/MF: 3.30V, 1A LDO, 8LD DFN Pkg.
- f) MCP1726-5002E/SN: 5.00V, 1A LDO, 8LD SOIC Pkg.
- g) MCP1726-ADJE/MF: Adjustable, 1A LDO, 8LD DFN Pkg.
- h) MCP1726-3002E/MF: Tape and Reel, 3.30V, 1A LDO, 8LD DFN Pkg.

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